

Optimizing Tri-Core Permanent-Magnet-Linear-Generator Direct-Drive Wave-Energy-Conversion System Design for Sea Wave Characteristics in South Coast Yogyakarta

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ABSTRACT

According to statistical data, the south coast Yogyakarta has significant ocean wave height which can be used to generate electricity by using wave-energy-converter system. One of the simplest way to convert wave energy to electricity is using direct-drive wave-energy-conversion (WEC) system with permanent-magnet-linear-generator (PMLG). This method is simple because it does not need to convert linear motion to rotational motion. However, PMLG has large electric power losses, has great weight in both of the stator and rotor, and expensive to make. In this paper, a tri-core PMLG was designed. The electric power losses in the winding, translator weight, and material cost were ideally minimized using multiobjective optimization combined with simulated annealing (SA) algorithm. Then, the design was verified using finite element analysis. The optimized design of this PMLG was simulated using sinusoidal ocean waves which usually occur in the south coast of Yogyakarta to analyze the performance of this linear generator. Simulation result has been shown that this generator can generate 911 watt peak output power at the rated condition and at the optimum load with 81.14% efficiency. This confirms that the optimized design of PMLG is suitable for direct-drive WEC with low power losses and material cost.

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1. INTRODUCTION

One of the world's renewable energy resources which available in a great amount is ocean waves energy. It is estimated that the total ocean waves energy in the world can be used to fulfill 2% from the world's total energy demand [1]. According to [2], south coast of Yogyakarta has potential to generate electric power about 1.9 MW. This is also supported based on the mapping result which conducts by Indonesian Agency for Meteorology Climatology and Geophysics that south coast of Yogyakarta has average ocean waves between 0.5 - 2 meters height [3]. It means that south coast of Yogyakarta has vertical velocity of the ocean waves about 0.5 m/s [4]. This velocity is able to make motion which can drive a generator.

There are several technology of wave-energy-conversion (WEC) system. One of the simple way to convert the ocean waves to electricity is using direct-drive WEC system technology [5]. Direct-drive WEC system is simple because it doesnot require to convert linear motion of the ocean waves to rotational motion to generate electricity, so the efficiency of the WEC is higher [6]. It happens because direct-drive WEC is using floater (buoy) coupled with linear generator to convert ocean waves motion into electricity, so that it is

only needs a few part to develop the WEC. Linear generator which is used to convert the linear motion of the ocean waves is classified based on the shape, that is: tubular, tri-core, square-core and tri-coil linear generator [7], [8]. Tri-core linear generator is cheaper, provide higher emf, and suitable with ocean waves between 0.5 until 1.5 meters height which is similar to the characteristic of ocean waves in south coast of Yogyakarta rather than the others linear generator topologies.

To obtain a good design of linear generator, an optimization procedure should be done to get global optimal solutions [9], [10]. Based on the literature review, there are still only a few research which related to the optimization of linear generator design, although the optimization procedure has been successfully used to optimize several electric machine, such as transformer, direct-current machine, synchronous machine, and the others electric machine [11-14]. Result was shown that optimization procedure could deliver better electric machine design rather than without optimization procedure.

This research proposes direct-drive WEC applied with tri-core permanent-magnet-linear-generator (PMLG) for converting ocean waves motion to electricity. To get a good design of PMLG, the optimization was performed in this linear generator. Three objectives function was used in this optimization, there are: electric power losses in the winding, translator weight, and material cost of the PMLG. That three objectives function then minimized using simulated annealing (SA) algorithm to get the best design. Then, the overall design of the PMLG was verified using finite element analysis (FEA) to get some parameters of the linear generator which could not be calculated analytically and to confirmed the magnetic parameter of the PMLG. Finally, the optimized design of tri-core PMLG was simulated using sinusoidal ocean waves characteristic to analyze and investigate the performance of the linear generator if the PMLG installed in south coast of Yogyakarta.

2. RESEARCH METHOD

2.1. Nomenclature

R_w internal resistance of AWG 11 (ohm/m), P_{dt} expected power of PMLG (watt), m number of phase, μ magnetic permeability, M_s number of armature, A_c area of the coil (m²), K_{cu} winding filling factor, D_w wire diameter (m), C_m air gap flux density coefficient, K_c Carter's coefficient, B_g air gap flux density (T), B_r permanent magnet residual flux (T), H_c magnetic flux coercive.

2.2. Site Selection for Installing Direct-Drive WEC

As mentioned above, the south coast of Java Island, especially in the south coast of Yogyakarta has large potential of ocean waves energy. According to the monthly data which has been collected by Indonesian Agency for Meteorology Climatology and Geophysics from 2000 until 2010, in this location, the significant wave height has average value of 1.44 meters. From that data, the WEC system was designed to work in rated condition with 1 meters wave height, so that the WEC could work well throughout the year.

$$v_s(t) = \omega \cdot \frac{H_s}{2} \cos(\omega t) \quad (1)$$

Using Equation (1), where angular velocity of the ocean wave (ω) was 1.76 rad/s, significant wave height (H_s) was 1 m, and ocean wave period was 3.55 s, so that the average vertical velocity of the ocean waves (v_s) was 0.6 m/s. It was assumed that the linear velocity of linear generator was the same as vertical velocity of the ocean waves since the direct-drive WEC became wave followers [15].

2.3. Mathematical Model of Tri-Core PMLG

Permanent-magnet-linear-generator (PMLG) with tri-core topology is one of the linear generator type which has triangular shape core in both of the stator and translator as shown in Figure 1(a). The design parameters of tri-core PMLG was shown in Figure 1(b). The aim of this shape was to simplify manufacturing process because of the simple shape and construction, so that it would minimize the manufacturing cost.

There are several mathematical models of tri-core PMLG which derived from the magnetical Equation of linear generator [16]. This mathematical models then optimized, so a good design of tri-core PMLG was obtained. The mathematical models are as follows:

- 1) Electrical current produced by PMLG (i_{ph}): from the expected power and emf produced by PMLG (E_{ph}), the Equation to determine the electrical current is given by

$$i_{ph} = \frac{P_{dt}}{mE_{ph}} \quad (2)$$

- 2) Winding resistance (R_{ph}): per phases winding resistance is influenced by internal resistance and length of wire (L_c), the Equation is as follows below

$$R_{ph} = R_w L_c N_{ph} \tag{3}$$

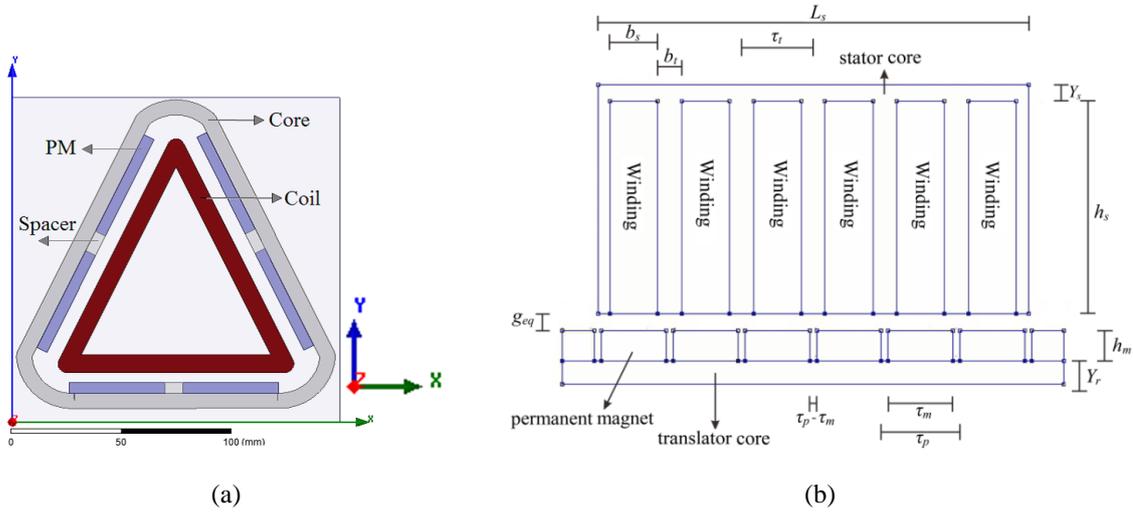


Figure 1. (a) Tri-core PMLG structure [8]; (b) Design parameters of tri-core PMLG [4]

where N_{ph} is number of turns per phase which is derived from number of turns per slot N_c . The Equation is shown by (4) and (5) respectively

$$N_{ph} = N_c p q \tag{4}$$

$$N_c = \frac{A_c K_{cu}}{\pi(D_w / 2)^2} \tag{5}$$

- 3) Geometry of tri-core PMLG: the geometry of tri-core PMLG will affect the phase resistance which also related to the electrical power losses in the winding, overall weight, and material cost needed to construct the linear generator. The mathematical geometry model of tri-core PMLG is mentioned in (6)-(11) with some of the nomenclature is mentioned in the next chapter.

$$A_c = h_s b_s \tag{6}$$

$$L_c = 2\pi(Y_s + \frac{h_s}{2}) + M_s W_s \tag{7}$$

$$K_c = \frac{\tau_t(5g + b_s)}{\tau_t(5g + b_s) - b_s} \tag{8}$$

$$g_{eq} = K_c \cdot g \tag{9}$$

$$h_m = \frac{g_{eq} B_g B_r}{\mu_0 |H_c| (B_r - B_g)} \tag{10}$$

$$\tau_m = C_m \tau_p \quad (11)$$

2.3. Optimization Procedure using Simulated Annealing

The design optimization of tri-core PMLG was done using multiobjectives function such as: electric power losses in the winding, translator weight, and material cost of the PMLG. Then, because the use of more than one single objectives function, so all of the objectives function must be transformed to a single objectives function using weighted sum approach. The method worked by using some weighted value $w_1, w_2, w_3, \dots, w_n$ with $w_1 + w_2 + w_3 + \dots + w_n = 1$. Each of the objective function was multiplied with the weighted value, the Equation for transforming the multiobjectives function to a single objectives function was shown below

$$F_t = \sum_{i=1}^n w_i F_i \quad (12)$$

where, w_i is the weighted value of the objectives function, F_i is each of single objectives function and F_t was the total single objectives function. Each of the objectives function was mentioned below:

- 1) Electrical power losses in the winding (P_{loss}): the electrical power losses in the winding was influenced by winding resistance, the Equation of the first objectives function was given by

$$F_1 = P_{loss} = i_p h^2 R_{ph} \quad (13)$$

- 2) Translator weight (m_{tr}): translator of the tri-core PMLG consisted of permanent magnet and translator core, to minimize the translator weight, the weight of permanent magnet and translator core must be minimized. The Equation was mentioned below respectively

$$F_2 = m_{tr} = m_{pm} + m_{tc} \quad (14)$$

where m_{pm} is permanent magnet weight and m_{tc} is the translator core weight. Each of the component could be calculated using Equation as mentioned below

$$m_{pm} = (\rho_m M_s h_m \tau_m W_s L_t) / \left(\tau_m + \frac{(\tau_p - \tau_m)}{2} \right) \quad (15)$$

$$m_{tc} = \rho_{tc} M_s Y_r L_t (W_s + 2(h_m + g_{eq})) \quad (16)$$

where ρ_m is the mass density of permanent magnet, ρ_{tc} is the mass density of translator core, and L_t is translator length which was designed to be 1 meters.

- 3) Material cost of the PMLG (c_{tot}): the material cost of tri-core PMLG consisted of: cost for both of translator and stator core, cost for permanent magnet, and cost for copper wire. Cost function could be calculated using weight of each part of the generator and the price per kilograms according to [17]. The Equation for calculating copper weight and stator core weight was mentioned below

$$m_{cu} = \rho_{cu} \pi (r_{wire})^2 L_c N_{ph} \quad (17)$$

$$m_{sc} = \rho_{sc} M_s W_s ((h_s + Y_s) L_s) - (b_s h_s p q m) \quad (18)$$

where ρ_{cu} is the mass density of copper wire, ρ_{sc} is the mass density of stator core. Then the material cost of tri-core PMLG could be calculated using Equation (10).

$$F_3 = c_{tot} = m_{pm} c_{pm} + (m_{tc} + m_{sc}) c_{fe} + m_{cu} c_{cu} \quad (19)$$

where c_{pm} is permanent magnet cost per kilogram, c_{fe} is steel cost per kilogram, and c_{cu} is copper cost per kilogram.

3. RESULTS AND ANALYSIS

3.1. Optimizing Tri-Core PMLG Design

It was decided to design a pico-scale tri-core PMLG with three-phase wye configuration and the expected output power is 1 kW. Several initial parameters were used in the design process, such as: average translator speed to be 0.6 m/s, the use of AWG 11 wire, and NdFeB permanent magnet to construct the linear generator. AWG 11 wire was chosen because the expected power of the generator was in pico-scale, so the current that flow through the winding was not exceed the current capability of AWG 11 which was 12 A. Then, NdFeB 52 MGOe was used because this type of permanent magnet has large flux density rather than the others type, so it would generate higher electromotive force (emf). With the same generated power, the higher emf would decrease the current generated from the PMLG, so it would decrease the electrical power losses in the winding. The NdFeB permanent magnet was mounted in the translator core surface.

Table 1. Specification of Tri-Core PMLG

| Variables | Symbol | Values | |
|---|------------|-------------------|------------------|
| | | Analytical Design | Optimized Design |
| Stator width (mm) | W_s | 50 | 50 |
| Number of slot per pole per phase | q | 1/3 | 1/3 |
| Number of poles | p | 12 | 12 |
| Average flux density in air gap (T) | B_{av} | 0.8 | 1 |
| Air gap (mm) | g | 2 | 0.4 |
| Flux density in the stator yoke (T) | B_{ys} | 1.8 | 2.2 |
| Flux density in the rotor yoke (T) | B_{yr} | 1.2 | 2 |
| Stator length (mm) | L_s | 148.68 | 313.4 |
| Pole-pitch (mm) | τ_p | 12.4 | 26.1 |
| Tooth-pitch (mm) | τ_t | 12.4 | 26.1 |
| Stator slot width (mm) | b_s | 8.26 | 17.41 |
| Stator tooth width (mm) | b_t | 4.1 | 8.7 |
| Real air gap (mm) | g_{eq} | 2.86 | 1 |
| Permanent magnet thickness (mm) | h_m | 5.17 | 3.66 |
| Permanent magnet length (mm) | τ_m | 11.2 | 23.5 |
| Stator yoke thickness (mm) | Y_s | 2.75 | 5.93 |
| Rotor yoke thickness (mm) | Y_r | 4.13 | 6.53 |
| Slot height (mm) | h_s | 217 | 60 |
| Number of stator turns per slot | N_c | 258 | 150 |
| Number of stator turns per phase | N_{ph} | 1031 | 600 |
| Average length of stator wire per turn (mm) | L_c | 849 | 375.8 |
| Weight of permanent magnet (kg) | m_{pm} | 0.8 | 1.188 |
| Weight of core (kg) | m_{fe} | 19.1 | 18.21 |
| Weight of copper (kg) | m_{cu} | 27.67 | 7.13 |
| Cost for permanent magnet (\$) | C_{pm} | 36.87 | 54.92 |
| Cost for core (\$) | C_{fe} | 75.3 | 72.12 |
| Cost for copper (\$) | C_{cu} | 365.3 | 94.21 |
| Winding resistance (ohm) | R_{ph} | 3.62 | 0.93 |
| Emf (volt rms) | E_{ph} | 58.34 | 42.52 |
| Max coil current (A) | i_{ph} | 5.7 | 7.84 |
| Electrical frequency (Hz) | f | 24.2 | 11.49 |
| Winding power losses (watt) | P_{loss} | 354.3 | 172.2 |
| Efficiency (%) | η | 64.57 | 82.78 |

Table 1 showed comparison between analytical and optimized design specification of tri-core PMLG. The optimized design of tri-core PMLG was delivered using simulated annealing algorithm combined with objective function which had been mentioned in (13)-(19) and several mathematical Equation of PMLG

design mentioned in (2)-(11). The weighting factor to optimize tri-core PMLG design had chosen to be 0.2 for weighting the electrical power losses objectives function, 0.7 for weighting the translator weight objectives function, and 0.1 for weighting the material cost objectives function. This weighting value was chosen because it could deliver the most optimum design of tri-core PMLG rather than the other weighting value. As seen in Table 1, the optimized design delivered lower winding resistance compared to the analytical design, it would give effect to the electrical power losses in the winding, so that the efficiency of the optimized design result of tri-core PMLG was higher compared to the analytical design result. Then, the translator weight delivered from optimization process was 10.16 kg, and it was less higher than the analytical result which gave 7.12 kg, it was due to the spreading of weighting value, so that all of the objectives function was in the optimum value. However, the overall weight of PMLG drawn from optimization result gave a great different. The overall weight from optimization result was 26.53 kg, but the analytical result gave 47.48 kg. This overall weight affected the material cost of tri-core PMLG, the material cost needed by optimized design was \$ 221.3, but the analytical result needed \$ 477.3 to build tri-core PMLG. It has been shown that optimization procedure successfully gives tri-core PMLG with optimum design specification.

3.2. Optimized Tri-Core PMLG Design Verification using FEA

The use of FEA was to analyze the two dimensional magnetic phenomenon in the optimized design, including to show unknown parameters which could not be determined using mathematical Equation such as winding inductance. It happened because FEA can deliver more accurate magnetical model result due to the use of differential Equation in magnetic field model. From the simulation result drawn from FEMM software, the maximum flux in the stator yoke was 1.83 T and the maximum flux in the translator yoke was 2.1 T. The magnetical flux in the translator yoke was larger than the assumption, but it was acceptable since the maximum flux only happened in a small area, so it did not make the translator yoke become excessive saturation. Another result delivered from FEMM showed that the phase inductance of tri-core PMLG was 0.049 H, this value was used to simulate and analyze the performance of tri-core PMLG in dynamic condition. Figure 2 shows flux density distribution in optimized tri-core PMLG design

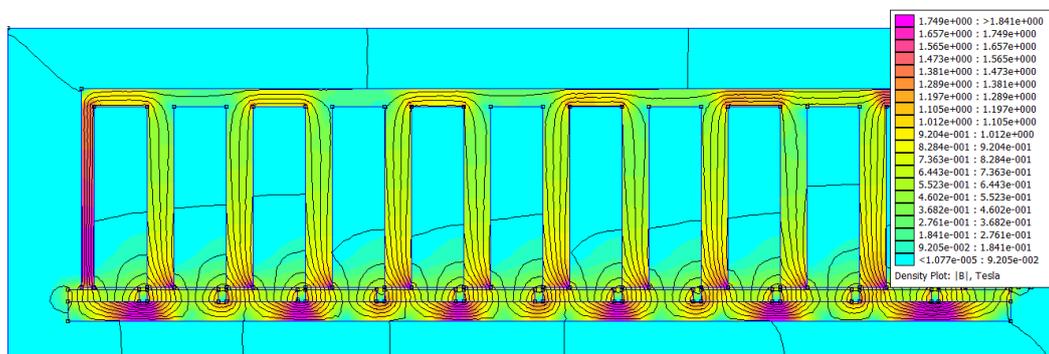


Figure 3. Flux density distribution in optimized tri-core PMLG design

3.3. Electrical Characteristic of PMLG

To analyze the performance of tri-core PMLG, a Simulink/MATLAB had been build. The first was to analyze the effect of load resistance to the output power and efficiency of tri-core PMLG. Simulation was conducted by varying load resistance of tri-core PMLG from 0 until 50 ohms with 1 ohm step size, and the translator speed was set to 0.6 m/s. Figure 4(a) showed the simulation result. As seen, the maximum output power was 589 watt, and it happened in 4 ohms load. The input power of the generator was 726 watt, so that since the mechanical power losses and core losses was neglected, the maximum efficiency of tri-core PMLG was 81.14%. This value was closest with the efficiency calculated in optimized design of PMLG.

When the optimal load was connected to the tri-core PMLG, the output characteristic of PMLG showed in Figure 4(b). It showed that the current was lagging to the emf produced by tri-core PMLG, it happened because of the phase inductance which present in the PMLG winding. However, the terminal voltage was in phase with the terminal current because of the resistive load. In the optimum load condition, the phase terminal voltage typically had a great different with the phase emf due to the resistance value of the PMLG. When the load resistance was increasing, the terminal voltage would also be increasing near the emf,

but when the load resistance was not in optimal value, the output power generated by PMLG was decreasing as shown in Figure 4(a).

In fact, the ocean waves never happened in a steady state condition. Usually the characteristic of the ocean waves was assume to be sinusoidal waves with a certain height and period. This condition made the translator speed changed sinusoidally, and it gave effect to the phase emf, phase current, and phase terminal voltage. The simulation result using sinusoidal ocean waves with constant height and period was shown in Figure 5(a). As seen, the peak magnitude of the emf changed sinusoidally according to the ocean waves movements. Since the vertical ocean wave speed changed sinusoidally, the input and output power of tri-core PMLG was represented in Figure 5(b). The peak input power of tri-core PMLG was 1,122 watt and the peak output power of tri-core PMLG was 911 watt. This value was near the expected value of initial design parameters of the generator. The input and output power of tri-core PMLG increased when the vertical ocean wave speed was increasing due to the increase of wave height. The effect of ocean wave height to the output power of tri-core PMLG was shown in Table 2. It has been shown that the generated output power of PMLG is intermittent since the ocean waves height is changed periodically.

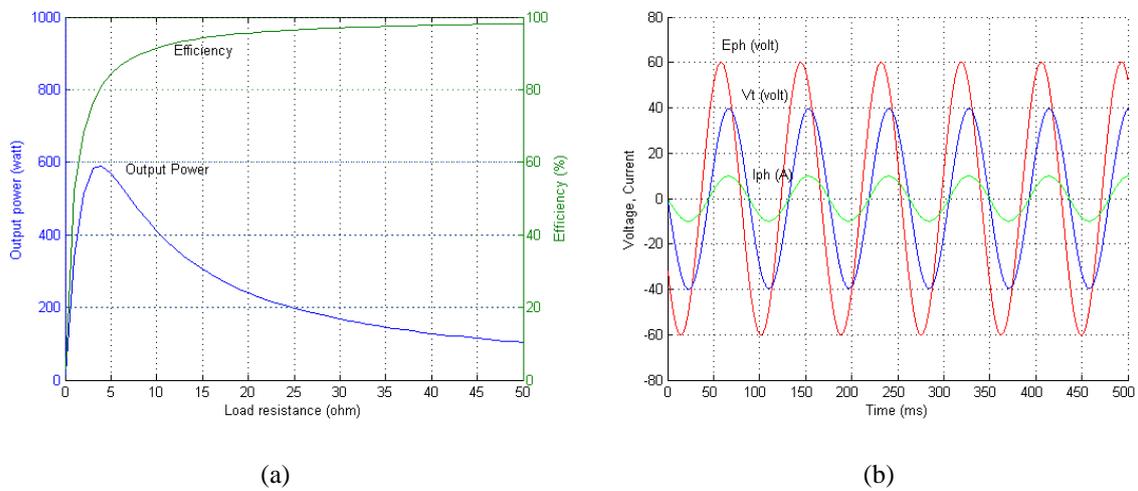


Figure 4. (a) Effect of load resistance to the output power and efficiency of tri-core PMLG; (b) Output characteristic of tri-core PMLG in steady state condition

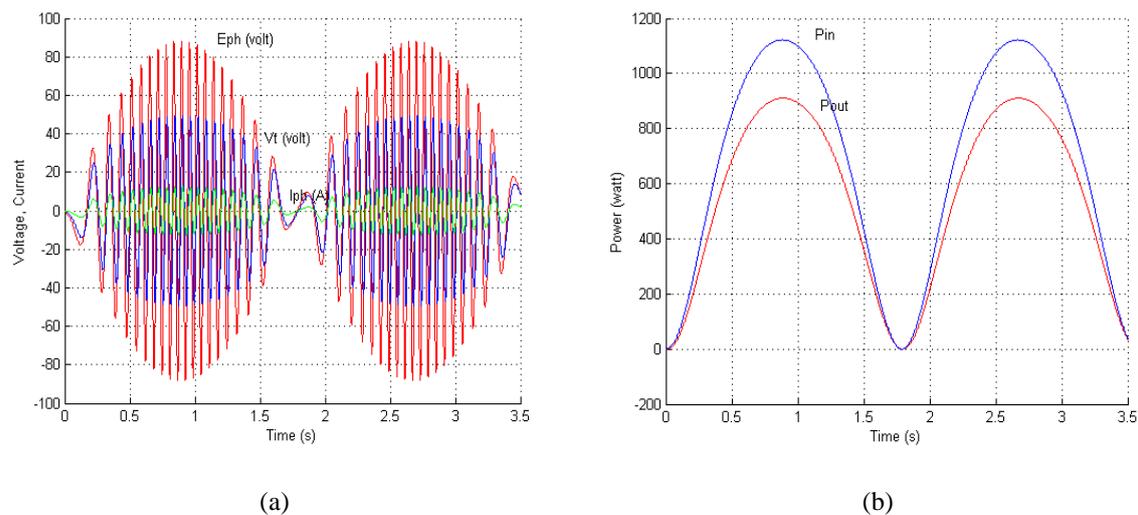


Figure 5. (a) Output characteristic of tri-core PMLG in sinusoidal ocean waves; (b) Input and output power of tri-core PMLG

Table 2. Effect of ocean wave height to the output of PMLG

| Wave height (m) | 0.2 | 0.5 | 1 | 1.5 | 2 | 2.5 | 3 |
|---------------------------------|-------|-------|-------|-------|-------|-------|--------|
| Peak output power (watt) | 310 | 628 | 911 | 1083 | 1197 | 1277 | 1334 |
| Peak phase emf (V) | 38.81 | 63.09 | 88.16 | 108.1 | 125.2 | 140.2 | 153.18 |
| Peak phase terminal volatge (V) | 28.59 | 40.9 | 49.22 | 53.72 | 56.46 | 58.33 | 59.57 |
| Peak phase current (A) | 7.15 | 10.23 | 12.3 | 13.43 | 14.11 | 14.58 | 14.89 |
| Frequency (Hz) | 5.75 | 7.66 | 11.49 | 13.4 | 15.33 | 17.24 | 19.16 |
| Peak power losses (W) | 72 | 145 | 211 | 252 | 278 | 297 | 313 |
| Efficiency (%) | 81.2 | 81.2 | 81.2 | 81.1 | 81.2 | 81.1 | 81.0 |

4. CONCLUSION

A small scale of wave-energy-conversion system has been designed and optimized using simulated annealing algorithm. From the optimization process, a good design of tri-core PMLG which has low electrical power losses in the winding, low translator weight, and low material cost has been obtained. The optimized design of tri core PMLG can generate 911 watt peak output power at the rated condition and at the optimum load with 81.14% efficiency. It has been shown since the translator is light in weight, the buoy will easily follows the ocean wave motion, it also happens in translator because the buoy and the translator is coupled. Furthermore, this condition will make the generator more easily to produce electrical power, rather than the PMLG which has heavy translator. Then, since the south coast of Yogyakarta is wavy and support the formation of waves between 0.5 to 1.5 meters height, the optimized design of PMLG is suitable for direct-drive WEC with low power losses and cost to install.

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Fransisco Danang Wijaya was born on February 1974 in Sleman, Indonesia. He received his Bachelor and Master degree, both from Electrical Engineering Major, Gadjah Mada University in 1997 and 2001 respectively. He then got his Doctor of Engineering in Tokyo Institute of Technology in 2009. He is currently Associate Professor at the Department of Electrical Engineering and Information Technology, Gadjah Mada University. His research is specialized in power system engineering, energy conversion, also transmission and distribution system. He is also expert in power system control technique using Magnetic Energy Recovery Switch (MERS).



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