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Optimum range of angle tracking radars: a theoretical computing

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ABSTRACT

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Keywords:

Angle error Angle tracking radars (ATRs) Optimal range Radar equation Signal-to-noise ratio (SNR) In this paper, we determine an optimal range for angle tracking radars (ATRs) based on evaluating the standard deviation of all kinds of errors in a tracking system. In the past, this optimal range has often been computed by the simulation of the total error components; however, we are going to introduce a closed form for this computation which allows us to obtain the optimal range directly. Thus, for this purpose, we firstly solve an optimization problem to achieve the closed form of the optimal range ($R_{opt.}$) and then, we compute it by doing a simple simulation. The results show that both theoretical and simulation-based computations are similar to each other.

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1. INTRODUCTION

There are different kinds of radar systems for different applications in industries, airports, military organs such as navy and so on. A type of radars is the tracking radars which are applied in fire control and etc [1]. Generally, two classes of tracking radars can be considered as angle tracking and range tracking radars. Our aim in this study attends to the first class. Figure 1 shows a real sample of tracking radar which has been placed on a ground platform for the fire control usages. Tracking radar for finding angle can use several tracking modes (scanning techniques) including sequential lobing, conical scan and monopulse (e.g. in terms of amplitude) [1]. These three techniques are different in terms of the error status. So, the mode used in each angle tracking radar must be considered in computing the error and then the optimal range of radar. Figure 2 shows a schematic which explains the mechanism of angle tracking radars. From step 1 to step 6, a target is seen in PPI display which tracked during these steps. Figure 3 and Figure 4 show the block diagram of two types of angle tracking radars based on monopulse and conical scan techniques, respectively [2], [3]. In [1]-[3], it is explained that in terms of angle error, two techniques of sequential lobing and conical scan are similar to each other. In the next parts, we specify the differences between the impacts of monopulse and conical scan/sequential lobing (non-monopulse) techniques. Computing the optimal range of each angle tracking radar is an optimization problem which depicts one of the main parameters of the radar. In [1], this problem has been represented, but it is only solved by simulation and numerical computation. As a completing approach for [1], we wish to solve this problem theoretically. In addition, we obtain the simulation result to compare to the proposed approach.

In the appendix of this paper, we give the readers a sample MATLAB code in order to the simulation-based computing of the optimal range.



Figure 1. A fire control tracking radar

This paper is organized as follows. In the second part, we have an overview on the resources of the angle error. In the third part, we theoretically solve the problem. In the forth part, we solve a typical example based on both simulation-based and theoretical approaches which allows us to have a comparison between them. The latest parts are allocated to the conclusions of the work and the appendix.



Figure 2. PPI display sequence to illustrate the target tracking process [2]

2. RESOURCES OF ERROR IN ATRS

According to [1], the general standard deviation of the angle error (σ_t) in ATRs is mainly depended on five components (five different standard deviations) including angle fluctuation error related to angle noise and glint phonomena (σ_g) , mechanical error $(\sigma_{mech.})$, control loop error $(\sigma_{cont.})$, scintillation error related to amplitude fluctuation and target fading $(\sigma_{sc.})$. And finally the error created by thermal noise of the radar system (σ_n) . As follows, we introduce the detailed information about σ_t . (1) shows the general formula of σ_t .

$$\sigma_t = \sqrt{\sigma_{cont.}^2 + \sigma_{mech.}^2 + \sigma_g^2 + \sigma_n^2 + \sigma_{sc.}^2}$$
(1)

Where $\sigma_{cont.}, \sigma_{mech.}, \sigma_g, \sigma_n$ and $\sigma_{sc.}$ ($\sigma_{sc.}$ is zero for monopulse mode) are as (2) to (6), respectively ((6) is for non-monopulse modes).

$$\sigma_{cont.} = \begin{cases} \frac{\omega_{\max}}{k_v} & \text{for fixed velocity} \\ \frac{\omega_{\max}}{k_a} & \text{for fixed acceleration} \end{cases}$$
(2)

$$\sigma_{mech.} = q_1 \theta_B$$
where $q_1 \in [0.001, 0.005]; q_1 t_{trained} = 0.002$
(3)

$$\sigma_g = \frac{L}{3R} \tag{4}$$

$$\sigma_{n} = \frac{\theta_{B}}{q_{2}\sqrt{(2 \times SNR)(\frac{PRF}{B_{s}})}}$$
(5)
where $q_{2} = \begin{cases} 1.3 & non - monopulse \\ 1.7 & monopulse \end{cases}$

$$\sigma_{sc} = 0.01\theta_B \tag{6}$$

monopulse

In the equations, B_s , k_v and k_a are some constants related to the status of movement of the target and depended on transform function of the control loop, see more information in [4]. w_{max} is the highest angle frequency of the target. θ_B is the maximum of beam width in two directions of elevation and azimuth $(\theta_B^{\circ} = \max\{\theta_{az}^{\circ}, \theta_{el}^{\circ}\})$. L is the largest size of the target and R is the distance between radar and target. SNR is signal-to-noise ratio of radar while receiving an echo from the target. SNR is based on the radar equation; we will discuss more about it later. PRF shows the pulse repetition frequency which is equal to the inverse of pulse repetition interval (shown by T_p) in pulse radar-based tracking systems. About SNR and radar equation, we can say that SNR is desired at the input side of the receiver and is therefore computable by the radar equation as (7); for simplicity, we omit some details of radar equation which are reachable in [1], [4].

$$SNR = \frac{P_t \lambda^2 G_a^2 RCS}{(4\pi)^3 P_{noise} R^4}$$
⁽⁷⁾

Where P_t the maximum of transmission power of the radar is, λ is the wavelength used in the radar. G_a is antenna gain of the radar (assume that we use a monostatic pulse radar for angle tracking) and described by (8) where $G_{o} = 32000$, and L_{el} and L_{az} are the size of the radar's antenna for elevation and azimuth, respectively, see more details in [1]. RCS shows the radar cross section of target and is not often fixed, especially in fast maneuvering and high rate fluctuation, however it is usually replaced with an average value which is at most equal to the geometrical cross section of the target (target size). So in here, we suppose that it is a constant. Finally, P_{noise} is power of the thermal noise of radar's receiver which is computed by (9); in this equation B is the noise bandwidth of the radar's receiver in which $B = \tau^{-1}$ (where τ is pulse width), F is the noise figure (NF), k and T_{o} are two constants; for more information in this regard, see [1].

$$G_a = \frac{G_{\circ} L_{el} L_{az}}{(60\lambda)^2}$$
(8)

$$P_{noise} = kT_{o}BF \tag{9}$$

In addition, there is another form for representing the noise power P_{noise} , refer to [1]. It is observable that in a specific example, all values of $P_t, \lambda, G_a, RCS, B, F$ and P_{noise} are fixed, thus the *SNR* is a function in terms of the radar range (*R*).



Figure 3. Block diagram of a monopulse tracking radar [2]



Figure 4. Block diagram of a conical scan radar [3]

3. PROPOSED METHOD

In this part, the proposed method is represented based on solving a mathematical optimization problem. In the previous part, we reviewed all details of angle error. It is noticeable that in an example with the specified parameters, σ_t is only in terms of R (distance to the target). Therefore, if we minimize the general standard deviation for a sample range such as $R = R_{opt.}$, then the total angle error becomes the minimum (because the mean of error variable is assumed to be zero, so a minimum variance/standard deviation shows the lowest error). In this case, $R_{opt.}$ is the optimal range. The equation of σ_t is rewritten as (10). If the SNR equation is to be rewritten as (11), we accordingly have σ_t as (12).

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$$\sigma_{t} = \left(\underbrace{\sigma_{cont.}^{2} + \sigma_{sc.}^{2} + (\frac{L}{3} \times \frac{1}{R})^{2} + (\frac{\left(\frac{\theta_{B}}{q_{2}}\right)^{2}}{\sqrt{K_{4}}} \times \frac{1}{SNR})}_{K_{2}} \times \frac{1}{SNR})\right)^{\frac{1}{2}}$$
(10)

$$SNR = \underbrace{\frac{P_t \lambda^2 G_a^2 RCS}{(4\pi)^3 P_{noise}}}_{K_3} \times \frac{1}{R^4}$$
(11)

$$\sigma_t = \left(K_1 + \frac{K_2}{K_3}R^4 + \frac{K_4}{R^2}\right)^{\frac{1}{2}}$$
(12)

In (12), all K_1 to K_4 coefficients are fixed, so we find the optimal range which can minimize the term for σ_t , as (13). After solving the optimization problem, we see that K_1 has no impact on the R_{opt} .

$$\frac{\partial \sigma_{t}}{\partial R} = 0 \rightarrow A.B = 0$$
where $A = \left(K_{1} + \frac{K_{4}}{R^{2}} + \frac{K_{2}}{K_{3}}R^{4}\right)^{-\frac{1}{2}}$

$$B = \frac{-2K_{4}}{R^{3}} + \frac{K_{2}}{K_{3}}(4R^{3})$$
(13)

The roots of (13) include all roots of two new equations of A = 0 and B = 0. It is provable that the equation A = 0 has no acceptable root for R (do it easily), we must solve the equation B = 0; (14) shows the single root of this equation.

$$R_{opt.} = \left(\frac{K_3 K_4}{2K_2}\right)^{\frac{1}{6}}$$
(14)

The achieved relation for the optimal range is a theoretical solution. This recent equation is a generally closed form for all scanning techniques due to the consideration of the coefficient q_2 (embedded in K_2) and the standard deviation of $\sigma_{sc.}$ in terms of different aspects related to the scanning techniques. In order to have a clear approach, we integrate all discussed items through the *Consequence* 1. Consequence 1. The optimal range of an angle tracking radar for all scanning techniques is as (15)-(18).

$$R_{opt.} = \sqrt[6]{a_1 a_2 a_3} \tag{15}$$

$$a_1 = \frac{P_t \lambda^2 G_a^2 RCS}{(4\pi)^3 P_{noise}}$$
(16)

$$a_2 = \frac{L^2}{9}$$
 (17)

$$a_3 = \frac{K^2 PRF}{\theta_B^2 B_s} \tag{18}$$

$$K \in \{1.3, 1.7\} \ (=q_2)$$

All angles are in degree(s) and other parameters are in linear space (not in dB) and some of them which have physical dimension are in S.I. units. In the next part, we solve an example based on the proposed approach and then compare it to the simulation result.

4. NUMERICAL EXAMPLE AND SIMULATION

In this part, we firstly use the MATLAB code in the appendix to solve the problem by simulation (classically). Its result is shown in Figure 5. It is clear that the optimal range when the simulation parameters are selected as Table 1 is between 5 and 10 meters.



Figure 5. The error curve shows the optimal range obtained from the simulation under the setting in Table 1

Table 1. Settings for simulation

$$\omega_{max} = 0.1 m$$

$$k_v = 10$$

$$\sigma_v = \omega_{max} / k_v = 0.001$$

$$\theta_B = 2^{\circ}$$

$$\sigma_{mech.} = 0.002 \theta_B = 0.004$$

$$L = 4 m$$

$$\sigma_g = \frac{L}{3R} = \frac{1.34}{R}$$

$$B_s = 30$$

$$PRF = 1000$$

$$P_t = 1000$$

$$P_{noise} = 1$$

$$\lambda = 0.03 m$$

$$G_a = 800$$

$$RCS = 12 m^2$$

$$K = 1.3$$

Now, we use the *Consequence* 1 for computing the optimal range theoretically. At first, a_1 , a_2 and a_3 are computed by (16) to (18).

$$a_1 = 3483.16$$

 $a_2 = 7.11$
 $a_2 = 14.08$
(19)

Then with using (15), we calculate the optimal range as follow.

$$R_{opt.} = \sqrt[6]{3483.16 \times 7.11 \times 14.08}$$

$$= 8.45 \ m \in [5,10]$$
(20)

Note that in theoretical computation of the optimal range, some of the parameters like ω_{max} , k_v and so on are not generally used (against the simulation-based solution, The MATLAB code for solving the problem based on simulation is in Table 2.).

5. CONCLUSIONS

In this work, our aim is to introduce a computational method [5]-[10] in which we theoretically computed the optimal range of ATRs. Our approach's result was matched with the simulation result in which they are sufficiently similar to each other. In fact, main aim of the work is to create an extension for the approaches represented in [1]. The proposed approach is suitable for educational aspects too.

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APPENDIX

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Table 2. The MATLAB code for solving the problem based on simulation
         sigma_t = zeros(1,1000);
         for R=1:1:1000
           w_max = 0.01;
           k_v = 10;
           sigma_v2 = (w_max/k_v).^2;
           theta_b = 2; % azimuth/elevation beam width
           sigma_mec2 = (0.002*theta_b).^2;
           L = 4; % it is the largest dimension of the tracked target.
           sigma_g2 = (0.33^*(L/R)).^2;
           Bs = 30;
           PRF = 1000;
           Pt = 1000;
           Lambda = 0.03;
           Ga = 800;
           RCS = 12;
           SNR = (Pt*(Lambda .^2)*(Ga.^2)*RCS)/((64*(pi^3))*(R.^4));
           % assume Pnoise = 1
           sigma_n2 = (theta_b/(1.3*sqrt(2*SNR*(PRF/Bs)))).^2;
           sigma_sc2 = (0.01*theta_b).^2;
           sigma_t(floor(R)) = sqrt (sigma_v2 +sigma_mec2 + sigma_g2
           + sigma_n2 + sigma_sc2);
         end
         figure;
         RR=1:1:1000;
         loglog(RR,sigma_t);grid;
         axis([0 100 10^(-1) 100]);
         title('Error Curve');
         xlabel('R (meter)');
         ylabel('General Standard Deviation of Angle Error');
```